

LTC3676/LTC3676-1

- ⁿ **Quad I2C Adjustable High Efficiency Step Down DC/DC Converters: 2.5A, 2.5A, 1.5A, 1.5A**
- Three 300mA LDO Regulators (Two Adjustable)
- \blacksquare **DDR Power Solution with V_{TT} and VTTR Reference**
- ⁿ **Pushbutton ON/OFF Control with System Reset**
- **n** Independent Enable Pin-Strap or I²C Sequencing
- ⁿ **Programmable Autonomous Power-Down Control**
- **Dynamic Voltage Scaling**
- Power Good and Reset Functions
- Selectable 2.25MHz or 1.12MHz Switching Frequency
- Always Alive 25mA LDO Regulator
- 12uA Standby Current
- **DED Low Profile 40-Lead 6mm** \times **6mm QFN and 48-Lead** Exposed Pad LQFP

APPLICATIONS

- Supports Freescale i.MX6, ARM Cortex, and Other Application Processors
- Handheld Instruments and Scanners
- \blacksquare Portable Industrial and Medical Devices
- Automotive Infotainment
- High End Consumer Devices
- Multi-Rail Systems

Power Management Solution for Application Processors

FEATURES DESCRIPTION

The [LTC®3676](http://www.linear.com/LTC3676) is a complete power management solution for advanced portable application processor-based systems. The device contains four synchronous step-down DC/DC converters for core, memory, I/O, and system on-chip (SoC) rails and three 300mA LDO regulators for low noise analog supplies. The LTC3676-1 has a \pm 1.5A buck regulator configured to support DDR termination plus a VTTR reference output. An I2C serial port is used to control regulator enables, power-down sequencing, output voltage levels, dynamic voltage scaling, operating modes and status reporting.

Regulator start-up is sequenced by connecting outputs to enable pins in the desired order or via the $1²C$ port. System power-on, power-off and reset functions are controlled by pushbutton interface, pin inputs, or 1^2C .

The LTC3676 supports i.MX, PXA and OMAP processors with eight independent rails at appropriate power levels. Other features include interface signals such as the VSTB pin that toggles between programmed run and standby output voltages on up to four rails simultaneously. The device is available in a 40-lead 6mm \times 6mm QFN and 48-lead exposed pad LQFP packages.

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TYPICAL APPLICATION

WAKE 1ms/DIV VDD(HIGH) VARM AND V_{SOC} V_{TT} AND V_{TTR} V_{LDO3} V_{DDR} 3676 TA01b 5V/DIV 1V/DIV 1V/DIV

Start-Up Sequence

3676fc

1

ABSOLUTE MAXIMUM RATINGS **(Note 1)**

PIN CONFIGURATION

ORDER INFORMATION

Consult LTC Marketing for parts specified with wider operating temperature ranges. *The temperature grade is identified by a label on the shipping container.

Consult LTC Marketing for information on nonstandard lead based finish parts.

For more information on lead free part marking, go to: http://www.linear.com/leadfree/

For more information on tape and reel specifications, go to: http://www.linear.com/tapeandreel/

junction temperature range, otherwise specifications are at T_A = 25°C (Note 2). V_{IN} = PV_{IN1} = PV_{IN2} = PV_{IN3} = PV_{IN4} = V_{IN_L2} = V_{IN_L3} = **VIN_L4 = DVDD = 3.8V. All regulators disabled unless otherwise noted.**

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Note 1: Stresses beyond those listed Under Absolute Maximum ratings may cause permanent damage to the device. Exposure to any Absolute Maximum rating condition for extended periods may affect device reliability and lifetime.

Note 2: The LTC3676 is tested under pulsed load conditions such that $T_J \approx T_A$. The LTC3676E is guaranteed to meet specifications from 0°C to 85°C junction temperature. Specifications over the –40°C to 125°C operating junction temperature range are assured by design, characterization and correlation with statistical process controls. The LTC3676I is guaranteed over the –40°C to 125°C operating junction temperature range and the LTC3676H is guaranteed over the full –40°C to 150°C operating junction temperature range. High junction temperatures

degrade operating lifetimes; operating lifetime is derated for junction temperatures greater than 125°C. The junction temperature $(T_J$ in °C) is calculated from the ambient temperature (T_A in \degree C) and power dissipation (PD, in Watts), and package to junction ambient thermal impedance $(\theta J_A$ in Watts/°C) according to the formula:

$$
T_J = T_A + (P_D \bullet \theta J_A).
$$

Note that the maximum ambient temperature consistent with these specifications is determined by specific operating conditions in conjunction with board layout, the rated package thermal impedance and other environmental factors.

7

ELECTRICAL CHARACTERISTICS

Note 3: The LTC3676 includes overtemperature protection that is intended to protect the device during momentary overload conditions. Junction temperature will exceed 150°C when overtemperature protection is active. Continuous operation above the specified maximum operating junction temperature may impair device reliability.

Note 4: Dropout voltage is defined as $(V_{1N} - V_{LDO1})$ for LDO1 or $(V_{IN_L} - V_{LDOX})$ for other LDOs when V_{LDOX} is 3% lower than V_{LDOX} measured with $V_{IN} = V_{IN-Lx} = 4.3V$.

Note 5: Dynamic supply current is higher due to the gate charge being delivered at the switching frequency.

Note 6: Soft-Start measured in test mode with regulator error amplifier in unity-gain mode.

Note 7: The switching regulator PMOS and NMOS on-resistance is guaranteed by correlation to wafer level measurements.

TYPICAL PERFORMANCE CHARACTERISTICS **VIN = 3.8V, TA = 25°C unless otherwise noted**

TYPICAL PERFORMANCE CHARACTERISTICS

9

3676 G13

TYPICAL PERFORMANCE CHARACTERISTICS

LDO2 to LDO4 Load Step Response

TEMPERATURE (°C)

0 50 100

150

3676 G20

–50

0

50

PIN FUNCTIONS **(QFN/LQFP)**

FB_L2 (Pin 1/Pin 2): Feedback Input for LDO2. Set fullscale output voltage using a resistor divider connected from LDO2 to this pin to ground.

V_{IN L2} (Pin 2/Pin 3): Power Input for LDO2. This pin should be bypassed to ground with a 1μF or greater ceramic capacitor. Voltage on V_{INL2} should not exceed voltage on V_{IN} pin.

LDO2 (Pin 3/Pin 4): Output Voltage of LDO2. Nominal output voltage is set with a resistor feedback divider that servos to a fixed 725mV reference. This pin must be bypassed to ground with a 1µF or greater ceramic capacitor.

LDO3 (Pin 4/Pin 5): Output Voltage of LDO3. Nominal output voltage is a fixed 1.8V. This pin must be bypassed to ground with a 1µF or greater ceramic capacitor.

V_{IN L3} (Pin 5/Pin 6): Power Input for LDO3. This pin should be bypassed to ground with a 1µF or greater ceramic capacitor. Voltage on V_{IN_L3} should not exceed voltage on V_{IN} pin.

LDO4 (Pin 6/Pin 7): Output Voltage of LDO4. Nominal output voltage is set with a resistor feedback divider that servos to a fixed 725mV reference. This pin must be bypassed to ground with a 1µF or greater ceramic capacitor.

V_{IN L4} (Pin 7/Pin 8): Power Input for LDO4. This pin should be bypassed to ground with a 1µF or greater ceramic capacitor. Voltage on $V_{\text{IN-14}}$ should not exceed voltage on V_{IN} pin.

FB_L4 (Pin 8/Pin 9): Feedback Input for LTC3676 LDO4. Set full-scale output voltage using a resistor divider connected from LDO4 to this pin to ground.

VDDQIN (Pin 8/Pin 9): V_{DD} Sense Input for LTC3676-1. Tie DDR memory V_{DD} supply to this pin.

EN_L4 (Pin 9/Pin 10): Enable LDO4 Input for LTC3676. Active high enables LDO4. A weak pull-down pulls EN_L4 low when left floating.

VTTR (Pin 9/Pin 10): DDR V_{RFF} Output Pin for LTC3676-1. Buffered reference equal to one-half VDDQIN voltage on Pin 8.

EN_L3 (Pin 10/Pin 11): Enable LDO3 Input. Active high enables LDO3. A weak pull-down pulls EN_L3 low when left floating.

SW4 (Pin 11/Pin 14): Switch Pin for Step-Down Switching Regulator 4. Connect one side of step-down switching regulator 4 inductor to this pin.

DV_{DD} (Pin 12/Pin 15): Supply Voltage for I²C Serial Port. This pin sets the logic reference level of SCL and SDA I²C pins. DV_{DD} resets ${}^{12}C$ registers to power-on state when driven to <1V. SCL and SDA logic levels are scaled to DV_{DD} . Connect a 0.1µF decoupling capacitor from this pin to ground.

SDA (Pin 13/Pin 16): Data Pin for the I²C Serial Port. The $1²C$ logic levels are scaled with respect to DV_{DD}.

SCL (Pin 14/Pin 17): Clock Pin for the I²C Serial Port. The I^2C logic levels are scaled with respect to DV_{DD}.

PV_{IN4} (Pin 15/Pin 18): Power Input for Step-Down Switching Regulator 4. Tie this pin to V_{IN} supply. This pin should be bypassed to ground with a 10μF or greater ceramic capacitor.

PV_{IN3} (Pin 16/Pin 19): Power Input for Step-Down Switching Regulator 3. Tie this pin to the V_{IN} supply. This pin should be bypassed to ground with a 10μF or greater ceramic capacitor.

EN_B4 (Pin 17/Pin 20): Enable Step-Down Switching Regulator 4. Active high input enables step-down switching regulator 4. A weak pull-down pulls EN_B4 low when left floating.

EN_B3 (Pin 18//Pin 21): Enable Step-Down Switching Regulator 3. Active high input enables step-down switching regulator 3. A weak pull-down pulls EN_B3 low when left floating.

VSTB (Pin 19/Pin 22): Voltage Standby. When VSTB is low, the DAC registers are selected by command register bit DVBxA[5]. When VSTB is high, the DAC registers are forced to DVBxB registers. Tie VSTB to ground if unused.

SW3 (Pin 20/Pin 23): Switch Pin for Step-Down Switching Regulator 3. Connect one side of step-down switching regulator 3 inductor to this pin.

PWR_ON (Pin 21/Pin 26): External Power On. Handshaking pin to acknowledge successful power-on sequence. PWR ON must be driven high within five seconds of WAKE going high to keep power on. PWR ON can be

11

PIN FUNCTIONS

used to activate the WAKE output by driving high. Drive low to shut down WAKE.

FB B3 (Pin 22/Pin 27): Feedback Input for Step-Down Switching Regulator 3. Set full-scale output voltage using resistor divider connected from the output of step-down switching regulator 3 to this pin to ground.

FB_B4 (Pin 23/Pin 28): Feedback Input for Step-Down Switching Regulator 4. Set full-scale output voltage using resistor divider connected from the output of step-down switching regulator 4 to this pin to ground.

FB B1 (Pin 24/Pin 29): Feedback Input for Step-Down Switching Regulator 1. Set full-scale output voltage using resistor divider connected from the output of step-down switching regulator 1 to this pin to ground.

FB_B2 (Pin 25/Pin 30): Feedback Input for Step-Down Switching Regulator 2. Set full-scale output voltage using resistor divider connected from the output of step-down switching regulator 2 to this pin to ground.

FB_L1 (Pin 26/Pin 31): Feedback Input for LDO1. Set output voltage using a resistor divider connected from LDO1 to this pin to ground.

VIN (Pin 27/Pin 32): Supply Voltage Input. This pin should be bypassed to ground with a 1µF or greater ceramic capacitor. All switching regulator PV_{IN} supplies should be tied to V_{IN}.

LDO1 (Pin 28/Pin 33): Always On LDO1 Output. This pin provides an always-on supply voltage useful for light loads such as a watchdog microprocessor or a real time clock. Connect a 1μF capacitor from LDO1 to ground.

ON (Pin 29/Pin 34): Pushbutton Input. A weak internal pull-up forces \overline{ON} high when left floating. A normally open pushbutton is connected from \overline{ON} to ground forcing a low state when pushed.

EN_L2 (Pin 30/Pin 35): Enable LDO2 Input. Active high enables LDO2. A weak pull-down pulls EN_L2 low when left floating.

SW2 (Pin 31/Pin 38): Switch Pin for Step-Down Switching Regulator 2. Connect one side of step-down switching regulator 2 inductor to this pin.

IRQ (Pin 32/Pin 39): Interrupt Request Output. Open-drain driver is pulled low for power good, undervoltage, and overtemperature warning and fault conditions. Clear IRQ by writing to the I²C CLIRQ command register.

WAKE (Pin 33/Pin 40): System Wake Up. Open-drain driver output releases high when signaled by pushbutton activation or PWR_ON input. It may be used to initiate a pin-strapped power-up sequence by connecting to a regulator enable pin.

EN_B2 (Pin 34/Pin 41): Enable Step-Down Switching Regulator 2. Active high input enables step-down switching regulator 2. A weak pull-down pulls EN B2 low when left floating.

PVIN2 (Pin 35/Pin 42): Power Input for Step-Down Switching Regulator 2. Tie this pin to V_{IN} supply. This pin should be bypassed to ground with a 10μF or greater ceramic capacitor.

PV_{IN1} (Pin 36/Pin 43): Power Input for Step-Down Switching Regulator 1. Tie this pin to V_{IN} supply. This pin should be bypassed to groundwith a 10μF or greater ceramic capacitor.

EN_B1 (Pin 37/Pin 44): Enable Step-Down Switching Regulator 1. Active high enables step-down switching regulator 1. A weak pull-down pulls EN_B1 low when left floating.

RSTO (Pin 38/Pin 45): Reset Output. Open-drain output pulls low when the always-on regulator LDO1 is below regulation or during a hard reset initiated by a pushbutton input or command registers.

PGOOD (Pin 39/Pin 46): Power Good Output. Open-drain output pulls low when any enabled regulator falls below power good threshold or during dynamic voltage slew unless disabled in command register. Pulls low when all regulators are disabled.

SW1 (Pin 40/Pin 47): Switch Pin for Step-Down Switching Regulator 1. Connect one side of step-down switching regulator 1 inductor to this pin.

GND (Exposed Pad Pin 41/Pin 49): Ground. The exposed pad must be connected to a continuous ground plane of the printed circuit board by multiple interconnect vias directly under the LTC3676 to maximize electrical and thermal conduction.

BLOCK DIAGRAM—LTC3676

3676fc

13

BLOCK DIAGRAM—LTC3676-1

INTRODUCTION

The LTC3676 is a complete power management solution for portable microprocessors and peripheral devices. It generates a total of eight voltage rails for supplying power to the processor core, DDR memory, I/O, always-on realtime clock and HDD functions. Supplying the voltage rails are an always-on low quiescent current 25mA LDO, two 2.5A step-down regulators, two 1.5A step-down regulators, and three 300mA low dropout regulators. Supporting the multiple regulators is a highly configurable power-on sequencing capability, dynamic voltage scaling DAC output voltage control, a pushbutton interface controller, control via an 1^2C interface, and extensive status and interrupt outputs.

The LTC3676-1 supports DDR memory applications by replacing the LTC3676 LDO4 feedback and enable pins with VDDQIN and VTTR pins. The DDR V_{DD} supply is connected to the LTC3676-1 VDDQIN pin. A buffered DDR termination voltage equal to one half the voltage on VD-DQIN is output on VTTR. The VTTR voltage is connected internally on the LTC3676-1 to the reference side of the Buck1 error amplifier. When Buck1 is configured with a gain of one, its output can be used as at DDR termination supply. Table 1 shows the functional differences between the LTC3676 and LTC3676-1.

Figure 1. V_{TT} Buck Regulator and VTTR Reference Block Diagram

Always-On 25mA Low Dropout Regulator

The LTC3676 includes a low quiescent current low dropout regulator that remains powered whenever a valid supply is present on V_{IN} . The always-on LDO1 remains active until V_{IN} drops below 2.0V (typical). This is below the 2.5V undervoltage thresholdineffectforthe restoftheLTC3676 circuits. The always-on LDO is used to provide power to a standbymicrocontroller, real-timeclock, orotherkeep-alive circuits. The LDO is guaranteed to support a 25mA load. A 1µF lowimpedanceceramicbypasscapacitorfromLDO1 to GND is required for compensation. A power good monitor pulls RSTO low whenever LDO1 is 8% below its regulation target. LDO1 has current limit circuitry to protect from short circuit and overloading. The output voltage of LDO1 is set with a resistor divider connected from LDO1 output pin to the feedback pin FB_L1, as shown in Figure 2. The output voltage is calculated using the following formula:

$$
V_{LDO1} = 725 \cdot \left(1 + \frac{R1}{R2}\right)(mV)
$$

300mA Low Dropout Regulators

Three LDO regulators on the LTC3676 will each deliver up to 300mA output. Each LDO regulator has separate input supply to help manage power loss in the LDO output devices. The LDO regulators are enabled by pin input or I2C command register. When disabled, the regulator outputs are pulled to ground through a 625 Ω resistor. A low ESR 1µF ceramic capacitor should be tied from the LDO output to ground. The 300mA LDO regulators have current limit control circuits. The LDO input voltages, V_{IN-L2} , V_{IN-L3} , and $V_{IN-1.4}$ must be at potential of V_{IN} or less.

The LDO regulator 1^2C command register controls are shown in Table 2 and Table 3.

LTC3676 Resistor Programmable LDO2 and LDO4

LDO2 and LDO4 output voltages are programmed by resistor dividers tied from the LDO output pin to the feedback pin as shown in Figure 2. The output voltage is calculated using the following formula:

$$
V_{LDO} = 725 \cdot \left(1 + \frac{R1}{R2}\right)(mV)
$$

Figure 2. LDO1, LDO2 and LDO4 Application Circuit

Fixed Output LDO3

Regulator LDO3 has a fixed voltage output of 1.8V.

*denotes default power-on value.

LDO4 Operation LTC3676-1

LDO4 on the LTC3676-1 has neither enable nor feedback pins. There are four LDO4 output voltages selectable by command register bits LDOB[4:3]. The power-on default output is 1.2V with selectable outputs of 2.5V, 2.8V, and 3.0V. LDO4 is enabled only through the command register bit LDOB[2].

LDO4 Command Register Controls

Table 3. LDO4 Control Command Register Settings

*denotes default power-on value.

STEP-DOWN SWITCHING REGULATORS

The LTC3676 contains four buck regulators. Two of the buck regulators are capable of delivering up to 2.5A load current and the other two can deliver up to 1.5A each. The regulators have forward and reverse current limiting, softstart, and switch slew rate control for lower radiated EMI.

The LTC3676 buck regulators are capable of 100% duty cycle, ordropout, regulation. Whenindropoutthe regulator output voltage is equal to PV_{IN} minus the load current times $R_{DS(ON)}$ of the converters PMOS device and inductor DCR.

Each buck regulator is enabled using its enable pin or ${}^{12}C$ command register control. Operating modes, start-up option, reference voltage, and switch slew rate are controlled using the I^2C port.

The buck converter $1²C$ command register controls are shown in Table 4, Table 5, Table 6, and Table 7.

Operating Modes

The buck regulators can operate in either pulse-skipping, Burst Mode operation, or forced continuous mode. In pulse-skipping setting the regulator will skip pulses at light loads but will operate at constant frequency. In Burst Mode setting the regulator operates in Burst Mode operation at light loads and in constant frequency PWM mode at higher load. In forced continuous setting the inductor current is allowed to be less than zero over the full range of duty cycles. In forced continuous operation the buck regulator has the ability to sink output current. Because the regulator is switching every cycle regardless of output load, forced continuous mode results in the least output voltage ripple at light load.

Output Voltage Programming

Each of the step-down converters uses a dynamically slewing DAC for its reference. The output voltage of the DAC reference is selectable using a 5-bit I2C command register. The output voltage is set by using a resistor divider connected from the step-down switching regulator output to its feedback pin as shown in Figure 3. The output voltage is calculated using the following formula:

$$
V_{OUT} = \left(1 + \frac{R1}{R2}\right) \cdot (DVBx \cdot 12.5 + 412.5) (mV)
$$

DVBx is the decimal value of the 5-bit binary number in the I²C command registers. The default DAC input code is 11001 (25 in decimal) which corresponds to a reference

voltage of 725mV. Typical values for R1 are in the range of 40k to 1M. Capacitor C_{FB} cancels the pole created by the feedback resistors and the input capacitance on the FB pin and helps to improve load step transient response. A value of 10pF is recommended.

Inductor Selection

The choice of step-down switching regulator inductor influences the efficiency and output voltage ripple of the converter. A larger inductor improves efficiency since the peak current is closer to the average output current. Larger inductors generally have higher series resistance that counters the efficiency advantage of reduced peak current.

Inductor ripple current is a function of switching frequency, inductance, V_{IN} , and V_{OIII} as shown in this equation:

$$
\Delta I_{L} = \frac{1}{f \cdot L} \cdot V_{OUT} \cdot \left(1 - \frac{V_{OUT}}{V_{IN}}\right)
$$

A good starting design point is to use an inductor that gives ripple equal to 30% output current. Select an inductor with a DC current rating at least 1.5 times larger than the maximum load current to ensure the inductor does not saturate.

Input and Output Capacitor Selection

Low ESR ceramic capacitors should be used at both the output and input supply of the switching regulators. Only X5R or X7R ceramic capacitors should be used since they have better temperature and voltage stability than other ceramic types.

Operating Frequency

The switching frequency of each of the LTC3676 switching regulators may be set using the I2C command registers. The default switching frequency is 2.25MHz and the selectable frequency is 1.125MHz. Operation at lower frequency improves efficiency by reducing internal gate charge and switching losses at the expense of a larger inductor.

The lowest duty cycle of the step-down converter is determined by minimum on-time. Minimum on-time is the shortest time duration that the converter can turn its top **Figure 3. Step-Down Switching Regulator Application Circuit** PMOS on and off again. The time is the sum of gate charge

time plus internal delays of the peak current sense and PWM control. If the converters duty cycle will be 20% or less at 2.25MHz it is recommended to use the 1.125MHz setting to avoid minimum duty cycle. If the duty cycle falls below the minimum on-time of the converter, the output voltage ripple will increase as the converter skips cycles.

The default setting for the LTC3676-1 Buck1 switching frequency is 1.125MHz to ensure minimum on time effects are avoided at DDR termination reference voltages.

Phase Selection

To reduce the cycle by cycle peak current drawn by the switching regulators, the clock phase at which each of the LTC3676 buck's PMOS switch turns on can be set using ¹²C command register settings.

Figure 4. Phase Settings Full- and Half-Speed Buck Clock

Switch Slew Rate Control

To help reduce EMI the switch rise time of each buck regulator is slew limited by default. A faster setting is selectable using the I²C buck command registers. The faster setting will improve efficiency if limited edge rate is not required.

Soft-Start

To reduce inrush current at start-up each buck regulator soft starts when enabled. When enabled the internal reference voltage is ramped from ground to the level of the slewing DAC output at a rate of 0.8V/ms. During soft-start the converter is forced to pulse-skipping mode regardless of command register mode settings.

*denotes default power on-value.

Table 5. Buck2 Control Command Register

*denotes default power-on value.

Table 6. Buck3 Control Command Register

*denotes default power-on value.

Table 7. Buck4 Control Command Register

*denotes default power-on value.

SLEWING DAC REFERENCE OPERATION

Each LTC3676 step-down switching regulators error amplifier reference voltage is supplied by a 5-bit DAC with an output voltage range of 412.5mV to 800mV in 12.5mV steps. One of two 5-bit codes stored in $1²C$ command registers is selected for input to the DAC. When a change in code is detected by the DAC control circuits, the output of the DAC is slewed at 3.5mV/µs to the new value.

Dynamic Voltage Scaling

Table 8 shows the command registers used to control dynamic voltage scaling (DVS) ofthe step-downswitching regulators input reference DAC. The command register bits DVB1A[4:0] and DVB1B[4:0] store two 5-bit inputs to the DAC reference for Buck1. The bit stored in command register DVB1A[5] selects either the 5 bits stored in DVB1A[4:0] or DVB1B[4:0] DAC as input to the DAC reference. Buck2, Buck3, and Buck4 operate the same way using their assigned "A" and "B" command registers shown in Table 8. When the DAC detects a change in its input code it automatically slews to the new value at a rate of 3.5mV/us. A DVS can be initiated using the I^2C select bit or using the VSTB pin.

The LTC3676 VSTB pin HIGH selects the 5 bits stored in all four DVBx "B" registers. This facilitates a simultaneous DAC slew between the values in the "A" registers and the values in the "B" registers. The VSTB pin is logically ORed with the 1^2C command register bit. If the 1^2C select bit is already set high, the "B" registers are already selected and VSTB will have no effect. If no change in output is desired using the VSTB pin, set the value in the "A" register equal to the value in the "B".

Command register bits DVB1B[5], DVB2B[5], DVB3B[5], and DVB4B[5] control whether the PGOOD status pin is pulled low while the DAC output is slewing. The default command register setting is to pull PGOOD pin low during DAC slew. During the DVS, PGOOD will be held low for just the duration of the DVS and the PGSTAT register is not affected.

Figure 5. Dynamic Voltage Scaling

19

Table 8. Buck1, Buck2, Buck3, and Buck4 Slewing DAC Control Command Registers

*denotes default power-on value.

PUSHBUTTON OPERATION

Operating Mode State Diagram

Figure 6 shows the state diagram of the LTC3676 enable and sequence controller. First application of power to V_{IN} pin brings the controller to the power-on reset/hard reset (POR/HRST) state. In this state the I^2C command registers have been set to their default values, only LDO1 is operating, and the device is waiting for pushbutton or PWR_ON inputs. Regulator enable pins and command register enable bits are ignored in POR/HRST state. In the POR/HRST state V_{IN} draws typically 12 μ A.

Figure 6. LTC3676 Operating Mode State Diagram

Power Up Using Pushbutton

When the \overline{ON} pin is held low for 400ms the WAKE pin is pulled high, enable pins are recognized, and the five second PWR_ON timer is started. If in the ON state and PWR_ON is low or a fault is detected, then WAKE is brought low and after a 1 second power-down time, the STANDBY state is entered. In STANDBY, the enable bits in the command registers are cleared and enable pins are ignored. Table 9 shows the control of command registers, enables, and WAKE at each state.

The 5 second power-on state is intended for the system to detect that power rails are correct and either drive PWR_ON pin high or set command register bit CNTRL[7] high to keep the rails active. If there were a system level problem

Figure 7. Power Up Using Pushbutton

keeping the processor from driving PWR_ON, then the LTC3676 will pull WAKE low, shut off all regulators, and enter the STANDBY state. The STANDBY state is also a low power, 12µA (typical) state.

Table 9. Register, Enable, WAKE Control During Operating Mode State Control

STATE	REGISTERS	ENABLES	WAKE
POR/HRST DEFAULT	R/W	Inhibited	LOW
5 SEC PWR_ON TIMER	R/W	Allowed	HIGH
0N	R/W	Allowed	HIGH
1 SEC OFF TIMER HRST	Set to POR Defaults	Sequence Down	LOW
1 SEC OFF TIMER STANDBY	² C Enable and SW Mode Bits Cleared	Sequence Down	LOW
STANDBY	R/W	Inhibited	LOW

Power Down Using Pushbutton

When in the ON state, the system controller is responsible for deciding what action to take when a pushbutton event occurs. By monitoring the \overline{IRQ} status pin and IRQSTATIO]

status register bit, the controller can detect a pushbutton request. If a power-down into standby state is desired then the controller should drive PWR_ON low and set command register bit CNTRL[7] low.

Button Status Indication

When a pushbutton pulls \overline{ON} low for 50ms in the ON state, \overline{IRQ} is pulled low and the PB status bit in the IRQSTAT[0] status register is set. \overline{IRQ} and the IRQSTAT status bit are active while \overline{ON} is low or for a minimum of 50ms.

Power Up and Down with PWR_ON

The PWR_ON pin is an alternative way to power up the LTC3676 instead of using the \overline{ON} pin. When PWR ON is driven high or command register CNTRL[7] is set high, WAKE is pulled HIGH and the LTC3676 passes through the 5 second PWR_ON timer to the ON state. Figure 9 shows PWR_ON and WAKE timing. WAKE stays high for a minimum of 5 seconds.

Figure 9. Power Up and Down with PWR_ON

POWER ON SEQUENCING

Enable Pin Operation

The LTC3676 enable pins facilitate pin-strapping output rails to enable pins to up-sequence the LTC3676 regulators in any order. Figure 10 shows an example of pin-strapped sequence connections. The enable pins normally have a 0.8V (typical) input voltage threshold.

If any enable is driven high, the remaining enable input thresholds switches to an accurate 400mV threshold. To ensure separation of the sequenced rails, there is a builtin 450µs delay from the enable pin threshold crossing to the internal enable of the regulator. Figure 11 shows the start-up timing of the example shown in Figure 10.

Figure 10. Pin-Strapped Power-On Sequence Application

Figure 11. Pin-Strapped Power-On Sequence

Software Control Mode

Once a power-up sequence is completed, each regulator may be enabled and disabled individually by the system as needed for power management requirements by using the command register bit CNTRL[5]. When CNTRL[5] is set high the regulators ignore the state of their enable pins and respond only to ¹²C command register bit settings. The software control mode bit is reset in the one second standby and hard reset timer states so a pin strapped sequence begins at the next LTC3676 power on.

Keep Alive Operation

Each regulator has a dedicated command register keep alive bit that, when set, forces a regulator to be enabled regardless of the enable pins, command register enable bits, or the operating state of the LTC3676. A hard reset or fault shutdown resets the keep alive bits.

POWER OFF SEQUENCING

Sequence down command registers SQD1 and SQD2 are used to set the time, relative to WAKE falling, that a regulator is disabled either by lowering PWR_ON, or a fault induced shutdown. Table 10 shows register settings for SQD1 and SQD2.

Table 10.Sequence Down Control Command Register Settings

COMMAND REGISTER[BIT]	VALUE	SETTING
SQD1[1:0]	$00*$ 01 10 11	Disable Buck1 at Falling WAKE Disable Buck1 at Falling WAKE + 100ms Disable Buck1 at Falling WAKE + 200ms Disable Buck1 at Falling WAKE + 300ms
SQD1[3:2]	$00*$ 01 10 11	Disable Buck2 at Falling WAKE Disable Buck2 at Falling WAKE + 100ms Disable Buck2 at Falling WAKE + 200ms Disable Buck2 at Falling WAKE + 300ms
SQD1[5:4]	$00*$ 01 10 11	Disable Buck3 at Falling WAKE Disable Buck3 at Falling WAKE + 100ms Disable Buck3 at Falling WAKE + 200ms Disable Buck3 at Falling WAKE + 300ms
SQD1[7:6]	$00*$ 01 10 11	Disable Buck4 at Falling WAKE Disable Buck4 at Falling WAKE + 100ms Disable Buck4 at Falling WAKE + 200ms Disable Buck4 at Falling WAKE + 300ms
SQD2[1:0]	$00*$ 01 10 11	Disable LDO2 at Falling WAKE Disable LD02 at Falling WAKE + 100ms Disable LD02 at Falling WAKE + 200ms Disable LDO2 at Falling WAKE + 300ms
SQD2[3:2]	$00*$ 01 10 11	Disable LDO3 at Falling WAKE Disable LD03 at Falling WAKE + 100ms Disable LD03 at Falling WAKE + 200ms Disable LDO3 at Falling WAKE + 300ms
SQD2[5:4]	$00*$ 01 10 11	Disable LDO4 at Falling WAKE Disable LD04 at Falling WAKE + 100ms Disable LD04 at Falling WAKE + 200ms Disable LD03 at Falling WAKE + 300ms

*denotes default power-on value.

Figure 12 shows an example of a shutdown sequence. In this example, the bits in command registers SQD1 and SQD2 are set so that LDO2, LDO3, and LDO4 shut off at the same time as WAKE. Buck2 and Buck4 shut off 100ms after WAKE. Buck3 shuts off 200ms after wake and Buck1 shuts off 300ms after WAKE.

FAULT DETECTION AND REPORTING

The LTC3676 has fault detection circuits that monitor for V_{IN} undervoltage, die overtemperature, and regulator output undervoltage. Status of the fault detect circuits is indicated by the IRQ and PGOOD pins and the IRQSTAT and PGSTAT status registers.

VIN Undervoltage

The undervoltage (UV) circuit monitors the input supply voltage, V_{IN} , and when the voltage falls below 2.45V creates a FAULT condition that forces the LTC3676 into the standby state. The LTC3676 also provides a (UV) warning that is triggered at user programmable V_{IN} voltages as shown in Table 11.

Table 11. Undervoltage Warning Threshold Command Register Settings

COMMAND REGISTER[BIT]	VALUE	FALLING V_{IN} THRESHOLD
CNTRL[4:2]	$000*$	2.7V
	001	2.8V
	010	2.9V
	011	3.0V
	100	3.1V
	101	3.2V
	110	3.3V
	111	3.4V

*denotes default power-on value.

Over Temperature

To prevent thermal damage the LTC3676 incorporates an overtemperature (OT) circuit. When the die temperature reaches 155°C the OT circuits create a FAULT condition that forces the LTC3676 into standby. When the OT circuit detects the temperature falls below 140°C the FAULT condition is cleared. The LTC3676 also has an OT warning circuit that indicates the die temperature is approaching the OT fault threshold. The OT warning threshold is user programmable as shown in Table 12.

Table 12. Overtemperature Warning Threshold Command Register Settings

*denotes default power-on value.

PGOOD Status Pin

The PGOOD open-drain status pin is pulled low when all regulatorsaredisabled. PGOODisreleasedwhenallenabled regulator outputs are above 93% of programmed value. When any enabled regulator output falls below 92% of its programmed value for longer than 50µs the PGOOD pin is pulled low. The 50µs transient filter on PGOOD prevents PGOOD glitches due to transients. If the error condition persists for longer than 20ms, the $\overline{\text{IRQ}}$ pin is pulled low and status register IRQSTAT bit 2 is set to indicate a persistent PGOOD fault. The PGOOD pin is held low for the duration of the low output condition plus 1ms. Figure 13 shows the timing of PGOOD during enable and fault events.

PGSTAT and MSKPG Registers

The power good status of each regulator is accessible through the LTC3676 12 C interface by reading the contents of the PGSTAT status register. Table 13 shows the PGSTAT register contents. The data in the PGSTATL register is held for the length of the low voltage condition plus 1ms. The data in the PGSTATRT register is held only for the duration of the low voltage condition.

Table 13. Power Good Status Register

Each regulator has a corresponding bit in the MSKPG status register as shown in Table 14. When set, a bit blocks the PGOOD pin from being pulled low in the event of a low output voltage fault from its matching regulator. Setting a bit in the MSKPG command register does not mask the status in the PGSTAT status register.

Table 14. Power Good Status Masking Command Register

*denotes default power-on value.

IRQ Status Pin

The \overline{IRQ} pin is pulled and latched low when undervoltage, overtemperature or persistent PGOOD events occur. The IRQ pin is cleared by addressing the CLIRQ command register or by holding \overline{ON} low for 50ms.

IRQSTAT and MSKIRQ Registers

The bits in the MSKIRQ command register are set to mask warning, fault, and pushbutton status reporting to the $\overline{\text{IRQ}}$ pin. When set to mask, the \overline{IRQ} pin is not pulled low as a result of a fault or warning. Even though the $\overline{\text{IRQ}}$ pin is not pulled low the masked bit is set in the IRQSTAT register. When undervoltage, overtemperature faults, and hard reset signals are masked, the $\overline{\text{IRQ}}$ pin is not pulled low but LTC3676 state controller is pushed into the STANDBY or POR/HRST state. Accessing the CLIRQ status register clears the latched bits in the IRQSTAT status register and releases the $\overline{\text{IRQ}}$ pin.

Table 16. Interrupt Request Mask Command Register

*denotes default power-on value.

IRQ and IRQSTAT are not cleared by hard reset or fault shutdown. If V_{IN} remains applied while the LTC3676 is in STANDBY or POR/HRST then IRQSTAT may be read on the subsequent power up to determine if a fault or hard reset occurred.

RSTO Status Pin

The LTC3676 RSTO status pin is pulled low when alwayson LDO1 is 8% below its programmed value or when the LTC3676 is in the one second HRST timer state.

Hard Reset

A hard reset can be initiated by holding the \overline{ON} pin low or writing to the HRST command register. Bit six of the CNTRL command register determines how long \overline{ON} must remain low to initiate the hard reset. A hard reset sets all 1^2C command register bits to their default power-on state. Table 17 shows the command register control of hard reset function.

*denotes default power-on value.

A hard reset command will push the LTC3676 state controller through the 1 second HRST timer state and into the POR/HRST state.

Fault Shutdown

An undervoltage or overtemperature fault will push the LTC3676 state controller through the 1 second standby timer state and into standby state. If a down sequence is selected in the command registers, it will be executed during the 1 second power down interval.

LTC3676-1 Operation

The LTC3676-1 option supports DDR memory operation by generating a DDR termination reference and supply rail equal to one-half the voltage applied to VDDQIN Pin 8.

An internal resistive divider creates a reference voltage of one-half the voltage on VDDQIN. This reference is used by the V_{TT} reference buffer to output one-half of VDDQIN on VTTR Pin 9. The VTTR voltage is used as the reference for 1.5A switching regulator 1 which is used as the DDR termination supply.

Figure 1 shows typical application connections for the LTC3676-1 DDR termination reference and termination supply.

LDO4 has 1^2C command register selectable output voltages of 1.2V (default), 2.5V, 2.8V and 3V and is enabled only using the $1²C$ command register. Table 18 shows the LDO4 command register controls for the LTC3676-1.

Table 18. LDO4 Control Command Register Setting (LTC3676-1)

*denotes default power-on value.

I 2C OPERATION

The LTC3676 communicates with a bus master using the standard $1²C$ 2-wire interface. The timing diagram in Figure 14 shows the relationship of the signals on the bus. The two bus lines, SDA and SCL must be high when the bus is not in use. External pull-up resistors or current sources, such as the LTC1694 SMBus accelerator, are required on SDA and SCL. The LTC3676 is both a slave receiver and slave transmitter. The 1^2C control signals, SDA and SCL are scaled internally to the DV_{DD} supply. DV_{DD} must be connected to the same power supply as the bus pull-up resistors.

The 1^2C port has an undervoltage lockout on the DV_{DD} pin. When DV_{DD} is below approximately 1V, the 1^2C serial port is cleared and the command registers are set to default POR values.

The complete I²C command register table is shown in Table 20.

I 2C Bus Speed

The 1^2C port operates at speeds up to 400kHz. It has built in timing delays to ensure correct operation when addressed from an $1²C$ compliant master device. It also contains input filters designed to suppress glitches should the bus become corrupted.

I 2C START and STOP Conditions

A bus master signals the beginning of communications by transmitting a START condition. A START condition is generated by transitioning SDA from HIGH to LOW while SCL is HIGH. The master may transmit either the slave write or the slave read address. Once data is written to the LTC3676, the master may transmit a STOP condition which commands the LTC3676 to act upon its new command set. A STOP condition is sent by the master by transitioning SDA from LOW to HIGH while SCL is HIGH. The bus is then free for communication with another l^2C device.

I 2C Byte Format

Each byte sent to or received from the LTC3676 must be 8 bits long followed by an extra clock cycle for the acknowledge bit. The data should be sent to the LTC3676 most significant bit (MSB) first.

I 2C Acknowledge

The acknowledge signal is used for handshaking between the master and the slave. When the LTC3676 is written to, it acknowledges its write address and subsequent data bytes. When it is read from, the LTC3676 acknowledges its read address only. The bus master should acknowledge data returned from the LTC3676.

An acknowledge generated by the LTC3676 lets the master know that the latest byte of information was received. The master generates the acknowledge related clock and releases the SDA line during the acknowledge clock cycle. The LTC3676 pulls down the SDA line during the write acknowledge clock pulse so that it is a stable LOW during the HIGH period of this clock pulse.

At the end of a byte of data transferred from the LTC3676 during a READ operation, the LTC3676 releases the SDA line to allow the master to acknowledge receipt of the data. Failure of the master to acknowledge data from the LTC3676 has no effect on the operation of the ${}^{12}C$ port.

Figure 14. LTC3676 I2C Serial Port Timing

I 2C Slave Address

The LTC3676 responds to factory programmed read and write addresses. The least significant bit of the address byte is 0 when writing data and 1 when reading data. Table 19 shows read and write addresses for the LTC3676 options.

Table 19. LTC3676 and LTC3676-1 I2C Read and Write Addresses

LTC PART NUMBER	R/\overline{W}	ADDRESS
LTC3676	₩	0111 1000, 0x78
LTC3676	R	0111 1001. 0x79
LTC3676-1	ℼ	0111 1010, 0x7A
LTC3676-1	R	0111 1011, 0x7B

I 2C Write Operation

The LTC3676 has twenty-two command registers for control input. They are accessed by the 1^2C port via a sub-addressed writing system.

A single write cycle of the LTC3676 consists of exactly three bytes except when a clear interrupt or hard reset command is written. The first byte is always the LTC3676 write address. The second byte represents the LTC3676 sub-address. The sub-address is a pointer which directs the subsequent data byte within the LTC3676. The third byte consists of the data to be written to the location pointed to by the sub-address.

As shown in Figure 15, the LTC3676 supports multiple sub-addressed write operations. Data pairs sent following the chip write address are interpreted as sub-address and data. Any number of sub-address and data pairs may be sent. The data in the command registers is not acted on by the LTC3676 until a STOP signal is issued.

The LTC3676 will keep interim writes to the registers when a repeat START condition occurs. A repeat start may be used to set up other devices on the $1²C$ bus prior to sending a STOP condition. The LTC3676 will act on the data written prior to the repeat start when a STOP condition is detected.

I 2C Read Operation

Figure 16 shows the LTC3676 command register read sequence. The bus master reads a byte of data from a LTC3676 command or status register by first writing the LTC3676 write address followed by the sub-address to be read from. The LTC3676 acknowledges each of the two bytes. Next, the bus master initiates a new START condition and sends the LTC3676 read address. Following the acknowledge of the read address by the LTC3676, the LTC3676 pushes data onto the I²C bus for the 8 clock cycles. The bus master then acknowledges the data on its ninth clock.

The last read sub-address that is written to the LTC3676 is stored. This allows repeated polling of a command or status register without the need to re-write its sub-address. Additionally, the last register written may be immediately read by issuing a START condition followed by read address and clocking out the data.

Table 20. LTC3676 Command Registers

Table 22. LTC3676 Status Registers

APPLICATIONS INFORMATION

THERMAL CONSIDERATIONS AND BOARD LAYOUT

Printed Circuit Board Power Dissipation

In order to ensure optimal performance and the ability to deliver maximum output power to any regulator, it is critical that the exposed ground pad on the backside of the LTC3676 package be soldered to a ground plane on the board. The exposed pad is the only GND connection for the LTC3676. Correctly soldered to a 2500mm2 ground plane on a double-sided 1oz copper board, the LTC3676 has a thermal resistance(θ_{IA}) of approximately 34°C/W. Failure to make good thermal contact between the exposed pad on the backside of the package and an adequately sized ground plane will result in thermal resistances far greater than 34°C/W. To ensure the junction temperature of the LTC3676 die does not exceed the maximum rated limit and to prevent overtemperature faults, the power output of the LTC3676 must be managed by the application. The total power dissipation in the LTC3676 is approximated by summing the power dissipation in each of the switching regulators and the LDO regulators. The power dissipation in a switching regulator is estimated by:

$$
P_{D(SWx)} = V_{OUTx} \cdot I_{OUTx} \cdot \frac{100 \text{-Eff\%}}{100} (W)
$$

Where V_{OUTX} is the programmed output voltage I_{OUTX} is the load current and Eff is the % efficiency that can be measured or looked up from the efficiency curves for the programmed output voltage.

The power dissipated by an LDO regulator is estimated by:

 $P_{D(LDOx)} = V_{IN(LDOx)} - V_{LDOx} \cdot I_{LDOx}$ (W)

where V_{LDOx} is the programmed output voltage, $V_{IN(LDOx)}$ is the LDO supply voltage, and I_{LD0x} is the output load current. If one of the switching regulator outputs is used as an LDO supply voltage, remember to include the LDO supply current in the switching regulator load current for calculating power loss.

An example using the equations above with the parameters in Table 23 shows an application that is at a junction temperature of 120°C at an ambient temperature of 55°C. LDO2, LDO3, and LDO4 are powered by step-down Buck2 and Buck4. The total load on Buck2 and Buck4 is the sum of the application load and the LDO load. This example is with the LDO regulators at one third rated current and the switching regulators at three quarters rated current.

	V _{IN}	V _{OUT}	APPLICATION LOAD (A)	TOTAL LOAD(A)	EFF (%)	P_D (mW)
LD ₀₁	3.8	1.2	0.01	0.010		26.00
LD ₀₂	1.8	1.2	0.1	0.100		60.00
LD ₀₃	3.3	1.8	0.1	0.100		150.00
LD ₀₄	3.3	2.5	0.1	0.100		80.00
Buck1	3.8	1.2	1.875	1.875	80	450.00
Buck2	3.8	1.8	1.775	1.875	85	506.25
Buck3	3.8	1.25	1.125	1.125	80	281.25
Buck4	3.8	3.3	0.925	1.125	90	371.25
Total Power =				1925		
Internal Junction Temperature at 55°C Ambient				120° C		

Table 23. LTC3676 Power Loss Example

Printed Circuit Board Layout

When laying out the printed circuit board, the following checklist should be followed to ensure proper operation of the LTC3676:

- 1. Connect the exposed pad of the package (Pin 41) directly to a large ground plane to minimize thermal and electrical impedance.
- 2. The switching regulator input supply traces to their decoupling capacitors should be as short as possible. Connect the GND side of the capacitors directly to the ground plane of the board. The decoupling capacitors provide the AC current to the internal power MOSFETs and their drivers. It is important to minimize inductance from the capacitors to the LTC3676 pins.
- 3. Minimize the switching power traces connecting SW1, SW2, SW3, and SW4 to the inductors to reduce radiated EMI and parasitic coupling. Keep sensitive nodes such as the feedback pins away from or shielded from the large voltage swings on the switching nodes.
- 4. Minimize the length of the connection between the step-down switching regulator inductors and the output capacitors. Connect the GND side of the output capacitors directly to the thermal ground plane of the board.

TYPICAL APPLICATIONS

LTC3676 PMIC Configured to Support Freescale i.MX6 Processor

TYPICAL APPLICATIONS

LTC3676-1 PMIC Configured to Support Freescale i.MX6 Processor with DDR V_{TT} and VTTR

PACKAGE DESCRIPTION

Please refer to http://www.linear.com/designtools/packaging/ for the most recent package drawings.

NOTE:

1. DRAWING IS A JEDEC PACKAGE OUTLINE VARIATION OF (WJJD-2)

2. DRAWING NOT TO SCALE

3. ALL DIMENSIONS ARE IN MILLIMETERS

4. DIMENSIONS OF EXPOSED PAD ON BOTTOM OF PACKAGE DO NOT INCLUDE

MOLD FLASH. MOLD FLASH, IF PRESENT, SHALL NOT EXCEED 0.20mm ON ANY SIDE, IF PRESENT

5. EXPOSED PAD SHALL BE SOLDER PLATED

6. SHADED AREA IS ONLY A REFERENCE FOR PIN 1 LOCATION ON THE TOP AND BOTTOM OF PACKAGE

PACKAGE DESCRIPTION

Please refer to http://www.linear.com/designtools/packaging/ for the most recent package drawings.

REVISION HISTORY

TYPICAL APPLICATION

 Sequenced Power for High Performance Processor and DDR Memory Using LTC3375 Parallelable Buck Converters

RELATED PARTS

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